

Chap. 9 Theory and Design of Ferrimagnetic Components

Basic Phenomena

9.1 Basic Properties of Ferrimagnetic Materials

9.2 Plane Wave Propagation in a Ferrite Medium

9.3 Propagation in a Ferrite-Loaded Rectangular Waveguide

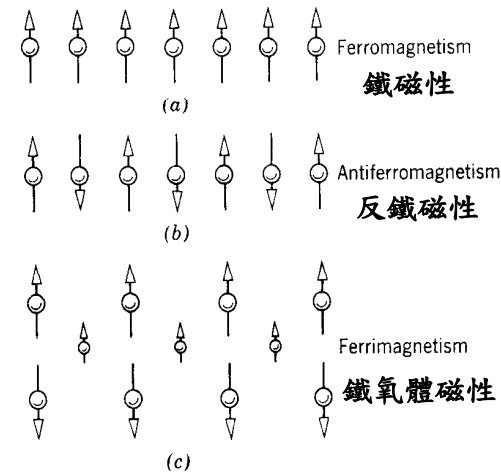
Applications

9.5 Ferrite Phase Shifters

9.6 Ferrite Circulator/Isolator

1

Introduction to ferrite materials



The ferrites are crystals having small electric conductivity compared to ferromagnetic materials,

Thus they are useful in high-frequency situations because of the absence of significant eddy current losses.

2

Properties of ferrite materials (I)

Nonreciprocal electrical property: the transmission coefficient through the device is not the same for different direction of propagation.

Unequal propagation constant: The left and right circularly polarized waves have different propagation constant along the direction of external magnetic field B_0 .

Anisotropic magnetic properties: The permeability of the ferrite is not a single scalar quantity, but instead is a tensor, which can be represented as a matrix.

3

Properties of ferrite materials (II)

Ferrites are **ceramiclike materials** with *specific resistivities* that may be as much as 10^{14} greater than that of metals and with *dielectric constants* around 10 to 15 or greater.

Ferrites are made by sintering **a mixture of metal oxides** and have the general chemical composition $MO \cdot Fe_2O_3$, where M is a divalent metal such as Mn, Mg, Fe, Zn, Ni, Cd, etc.

Relative permeabilities of **several thousand** are common. The magnetic properties of ferrites arise mainly from the **magnetic dipole moment** associated with the **electron spin**.

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Classical picture of the magnetization process

--- By treating the spinning electron as a gyroscopic top.

If an electron is located in a uniform static magnetic field \mathbf{B}_0 , a torque is given by

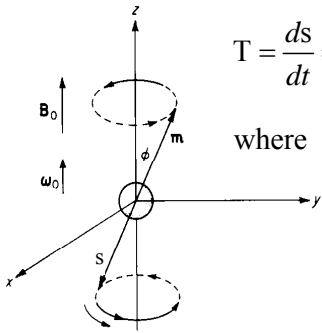
$$\mathbf{T} = \mathbf{m} \times \mathbf{B}_0 = -\mu_0 \gamma \mathbf{s} \times \mathbf{H}_0$$

$$\mathbf{T} = \frac{ds}{dt} = \frac{-1}{\gamma} \frac{dm}{dt} \Rightarrow \frac{dm}{dt} = -\mu_0 \gamma \mathbf{m} \times \mathbf{H}_0$$

where $\omega_0 = \frac{eB_0}{m_e}$ is called the Larmor frequency;

$s = \frac{\hbar}{2}$ is spin angular momentum;

$m = \frac{e}{m_e} \frac{\hbar}{2} = -\gamma s$ is magnetic dipole moment.



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Quantum mechanics' viewpoint $s_z = \pm 1/2$

In the absence of any damping forces, the actual precession angle will be determined by the initial position of the magnetic dipole, and the dipole will precess about \mathbf{B}_0 at this angle indifferently (free precession).

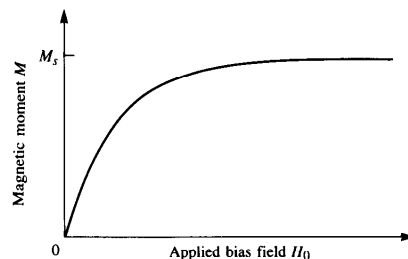
In reality, however, the existence of damping forces will cause the magnetic dipole to *spiral in* from its initial angle until \mathbf{m} is aligned with \mathbf{B}_0 .

This explains why s_z equals $\pm 1/2$ in the Quantum Mechanics.

But where does the damping force come from?

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Saturation magnetization



As the strength of the bias field H_0 is increased, more magnetic dipole moments will align with H_0 until all are aligned, and \mathbf{M} reaches an upper limit.

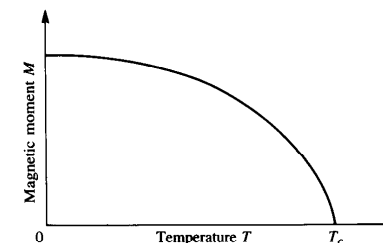
The material is then said to be magnetically saturated, and M_s is denoted as the saturation magnetization. M_s typically ranges from $4\pi M_s = 300$ to 5000 Gauss.

Below saturation, ferrite materials can be very lossy at microwave frequencies, and rf interaction is reduced.

The ferrites are usually operated in the saturated state.

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Curie temperature



The saturation magnetization of a material is a strong function of temperature, decreasing as temperature increases.

This effect can be understood by noting that the vibrational energy of an atom increases with temperature, making it more difficult to align all the magnetic dipoles.

At a high enough temperature a zero net magnetization results. This temperature is called the Curie temperature, T_c .

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Properties of some ferrite materials

Material	Trans-Tech Number	$4\pi M_s$ G	ΔH Oe	ϵ_r	$\tan \delta$	T_c °C	$4\pi M_r$ G
Magnesium ferrite	TT1-105	1750	225	12.2	0.00025	225	1220
Magnesium ferrite	TT1-390	2150	540	12.7	0.00025	320	1288
Magnesium ferrite	TT1-3000	3000	190	12.9	0.0005	240	2000
Nickel ferrite	TT2-101	3000	350	12.8	0.0025	585	1853
Nickel ferrite	TT2-113	500	150	9.0	0.0008	120	140
Nickel ferrite	TT2-125	2100	460	12.6	0.001	560	1426
Lithium ferrite	TT73-1700	1700	<400	16.1	0.0025	460	1139
Lithium ferrite	TT73-2200	2200	<450	15.8	0.0025	520	1474
Yttrium garnet	G-113	1780	45	15.0	0.0002	280	1277
Aluminum garnet	G-610	680	40	14.5	0.0002	185	515

Why use $4\pi M_s$? $\mathbf{B} = 4\pi\mathbf{M} + \mathbf{H} = \mu\mathbf{H}$ (Gaussian unit)

The unit of \mathbf{B} is Gauss; the unit of \mathbf{H} is Oersted. They have same dimension.

What does ΔH and M_r mean?

Ferrite linewidth and remanent magnetization

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HIMAG & Trans-Tech特性比較

• 高科磁技 HIMAG

Item	Unit	Measuring Condition	MZ-30	MZ-2	MZ-4	MZ-5	M 07
Initial Permeability	μ_i	100KHz 0.5mA	3000 ± 25%	2500 ± 25%	2400 ± 25%	5000 ± 25%	7000 ± 25%
Core Loss	Pc	mW/c.c. 25KHz sine wave 200mT	25	160	150		
			60	120	120		
			100	150	115		
			120	180	120		
			25		800	720	
Relative Loss Factor	tan δ / μ_i (*10-6)	10KHz 0.5mA 100KHz 0.5mA	60		650	550	
			100		650	480	
			120		750	600	
Saturation Flux Density	Bms	mT	430	450	430	400	400
Remanence	Brms	mT	150	150	150	130	130
Coercivity	Hc	Oe	0.15	0.15	0.15	0.12	0.1
Curie temperature	Tc		>190	>210	>210	>130	>120

• Trans-Tech

	Saturation Magnetization $4\pi M_s$	Line Width Factor g-eff	Line Width ΔH on ϕ 3dB	Dielectric Constant E'	Dielectric Loss Tangent TanE"/E'	Curie Temperature T _c (°C)	Spin Wave Line Width ΔH_{sw} oe	Remanent Induction B _r (Gauss)	Coercive Force H _c (oe)	Initial Permeability μ_0
TTVG-800	800 ± 5%	2	≤15	13.9 ± 5%	<0.002	182	2.0	580	80%	128
TTVG-930	930 ± 5%	2	≤10	14.0 ± 5%	<0.002	188	2.0	380	80%	225
TTVG-1000	1000 ± 5%	2	≤10	14.0 ± 5%	<0.002	199	2.0	320	30%	210
TTVG-1100	1100 ± 5%	2	≤10	14.1 ± 5%	<0.002	205	2.0	800	80%	209

Anisotropic magnetic properties (I)

If \bar{H} is the applied ac field, the total magnetic field is $\bar{H}_t = H_0\hat{z} + \bar{H}$,

where $|\bar{H}| \ll H_0$. The field produced a total magnetization on the ferrite is given by $\bar{M}_t = M_s\hat{z} + \bar{M}$

M_s is the dc saturation magnetization and \bar{M} is the additional ac magnetization (in the xy plane) caused by applied field.

$$\left. \begin{aligned} \bar{M}_t &= M_s\hat{z} + \bar{M} \\ \bar{H}_t &= H_0\hat{z} + \bar{H} \end{aligned} \right\} \Rightarrow \frac{d\bar{M}}{dt} = -\mu_0\gamma\bar{M} \times H_0$$

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Anisotropic magnetic properties (I)

The component equations of motion:

$$\frac{dM_x}{dt} = -\mu_0\gamma M_y(H_0 + H_z) + \mu_0\gamma(M_s + M_z)H_y$$

$$\frac{dM_y}{dt} = -\mu_0\gamma M_x(H_0 + H_z) - \mu_0\gamma(M_s + M_z)H_x$$

$$\frac{dM_z}{dt} = -\mu_0\gamma M_xH_y + \mu_0\gamma M_yH_x$$

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Anisotropic magnetic properties (II)

Omitting higher order terms, the equations can be reduced to

$$\left. \begin{aligned} \frac{dM_x}{dt} &= -\omega_0 M_y + \omega_m H_y, \\ \frac{dM_y}{dt} &= -(\omega_0 M_x + \omega_m H_x), \\ \frac{dM_z}{dt} &= 0, \text{ where } \omega_0 = \mu_0 \gamma H_0 \text{ and } \omega_m = \mu_0 \gamma M_s \end{aligned} \right\} \Rightarrow \left. \begin{aligned} \frac{d^2 M_x}{dt^2} + \omega_0^2 M_x &= \omega_m \frac{dH_y}{dt} + \omega_0 \omega_m H_x, \\ \frac{d^2 M_y}{dt^2} + \omega_0^2 M_y &= -\omega_m \frac{dH_x}{dt} + \omega_0 \omega_m H_y. \end{aligned} \right.$$

If \bar{M} and $\bar{H} \propto e^{j\omega t}$, the above equations can be reduced to the phasor equations:

$$\begin{aligned} (\omega_0^2 - \omega^2)M_x &= \omega_0 \omega_m H_x + j\omega \omega_m H_y, \\ (\omega_0^2 - \omega^2)M_y &= -j\omega \omega_m H_x + \omega_0 \omega_m H_y. \end{aligned} \Rightarrow \bar{M} = [\chi] \bar{H} = \begin{bmatrix} \chi_{xx} & \chi_{xy} & 0 \\ \chi_{yx} & \chi_{yy} & 0 \\ 0 & 0 & 0 \end{bmatrix} \bar{H},$$

where $\chi_{xx} = \chi_{yy} = \frac{\omega_0 \omega_m}{\omega_0^2 - \omega^2}$ and $\chi_{xy} = -\chi_{yx} = \frac{j\omega \omega_m}{\omega_0^2 - \omega^2}$

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Anisotropic magnetic properties (III)

To relate B and H , we have

$$\bar{B} = \mu_0(\bar{M} + \bar{H}) = [\mu] \bar{H} \Rightarrow [\mu] = \mu_0([U] + [\chi]) = \begin{bmatrix} \mu & j\kappa & 0 \\ -j\kappa & \mu & 0 \\ 0 & 0 & \mu_0 \end{bmatrix}$$

$$\left\{ \begin{aligned} \mu &= \mu_0(1 + \chi_{xx}) = \mu_0(1 + \chi_{yy}) = \mu_0 \left(1 + \frac{\omega_0 \omega_m}{\omega_0^2 - \omega^2}\right) \\ \kappa &= -j\mu_0 \chi_{xy} = j\mu_0 \chi_{yx} = \mu_0 \frac{\omega \omega_m}{\omega_0^2 - \omega^2} \end{aligned} \right.$$

A material having a permeability tensor of this form is called gyrotropic.

How to apply this concept to a circularly polarized wave?

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Forced precession of spinning electron (I)

If a small ac magnetic field is superimposed on the static field \mathbf{H}_0 , the magnetic dipole moment will undergo a forced precession.

Of particular interest is the case where the ac magnetic field is circularly polarized in the plane perpendicular to \mathbf{H}_0 .

A right-hand circularly polarized wave can be expressed in phasor form as

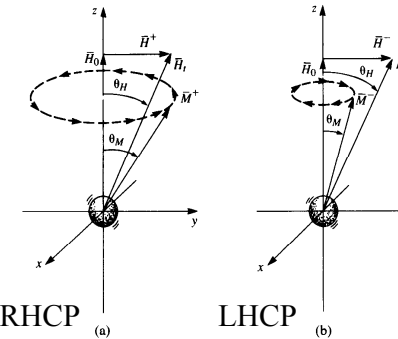
$$\bar{H}^+ = H^+(\hat{x} - j\hat{y})$$

and in time-domain form as

$$\bar{H}^+ = \text{Re}\{\bar{H}^+ e^{j\omega t}\} = H^+(\hat{x} \cos \omega t + \hat{y} \sin \omega t)$$

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Forced precession of spinning electron (II)



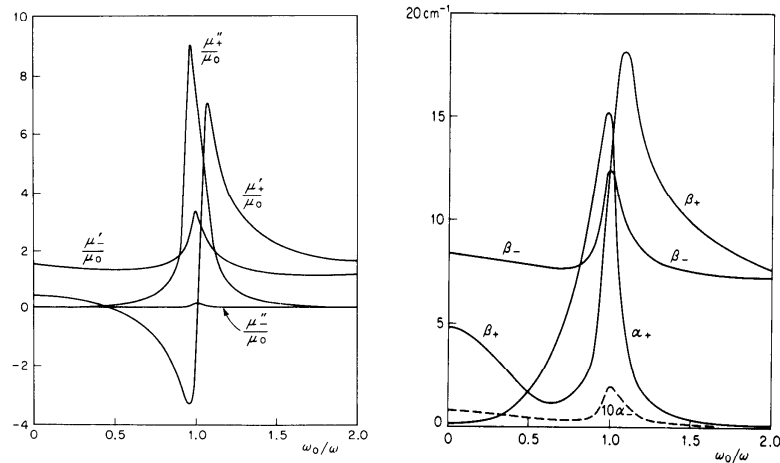
propagation constant

$$\beta_{\pm} = \omega \sqrt{\epsilon \mu^{\pm}}$$

$$\left. \begin{aligned} M_x^+ &= \frac{\omega_m}{\omega_0 - \omega} H^+, \\ M_y^+ &= \frac{-j\omega_m}{\omega_0 - \omega} H^+. \end{aligned} \right\} \Rightarrow \bar{M}^+ = \frac{\omega_m}{\omega_0 - \omega} \bar{H}^+ \Rightarrow \left\{ \begin{aligned} \mu^+ &= \mu_0 \left(1 + \frac{\omega_m}{\omega_0 - \omega}\right) \text{ RHCP} \\ \mu^- &= \mu_0 \left(1 + \frac{\omega_m}{\omega_0 + \omega}\right) \text{ LHCP} \end{aligned} \right.$$

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Real and imaginary permeability & propagation and attenuation constant.

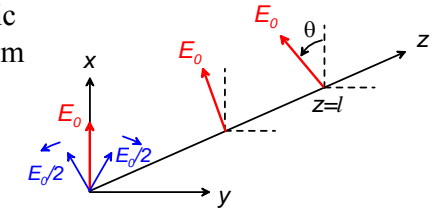


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Faraday rotation --- a nonreciprocal effect

Consider linearly polarized electric field at $z = 0$, represented as the sum of a RHCP and a LHCP wave:

$$\vec{E}|_{(z=0)} = \hat{x}E_0 = \frac{E_0}{2}(\hat{x} - j\hat{y}) + \frac{E_0}{2}(\hat{x} + j\hat{y})$$



These two polarized waves propagate with different propagation constants.

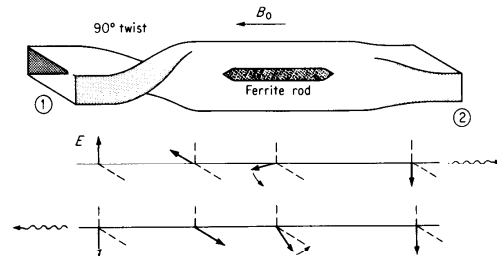
$$\begin{aligned}\vec{E}|_{(z=l)} &= \frac{E_0}{2}(\hat{x} - j\hat{y})e^{-j\beta_+l} + \frac{E_0}{2}(\hat{x} + j\hat{y})e^{-j\beta_-l} \\ &= E_0 \left[\hat{x} \cos\left(\frac{\beta_+ - \beta_-}{2}l\right) - \hat{y} \sin\left(\frac{\beta_+ - \beta_-}{2}l\right) \right] e^{-j(\beta_+ + \beta_-)l/2}\end{aligned}$$

$$\theta = \tan^{-1} \frac{E_y}{E_x} = -\left(\frac{\beta_+ - \beta_-}{2}\right)l. \quad \text{This effect is called Faraday rotation.}$$

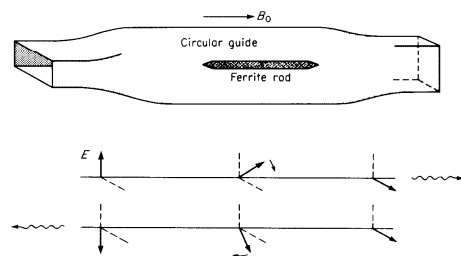
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Microwave gyrator

Gyrator with a twist section.

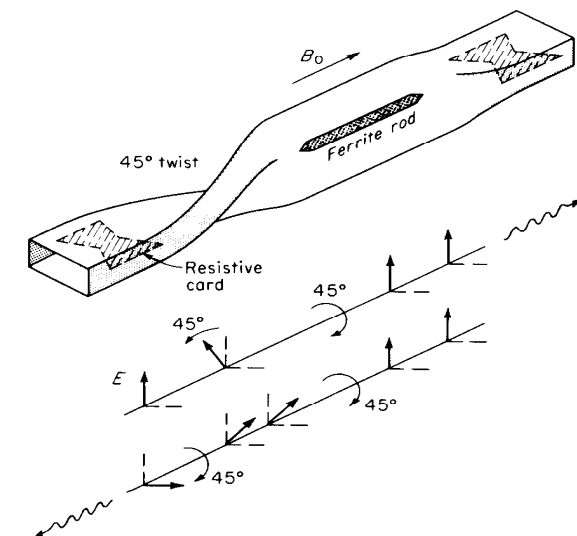


Gyrator without a twist section.



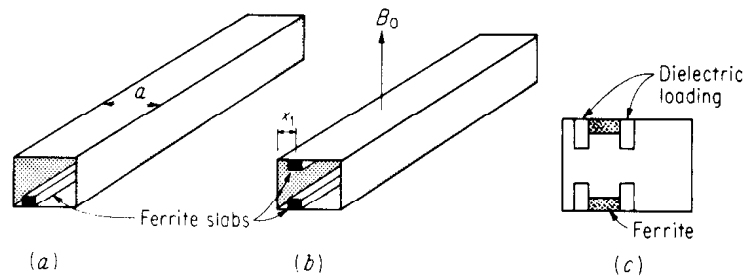
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Faraday-rotation isolator



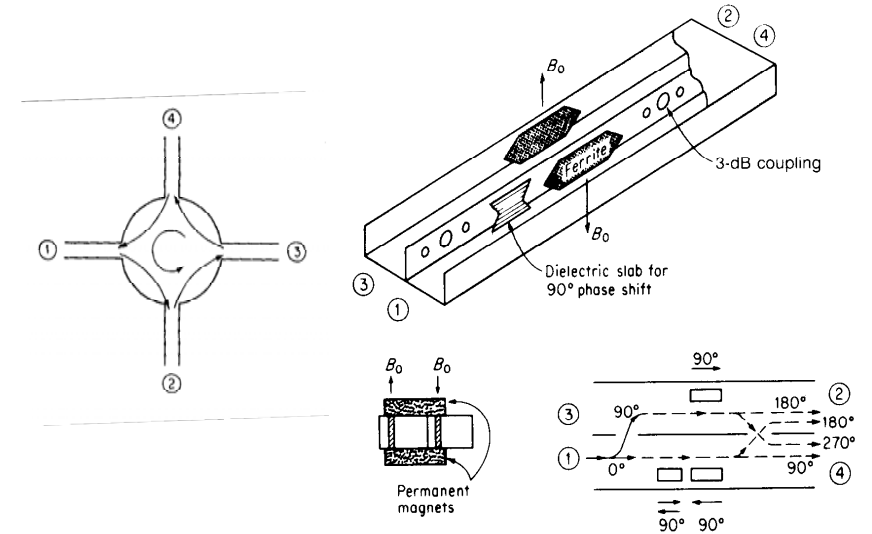
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Resonance isolator



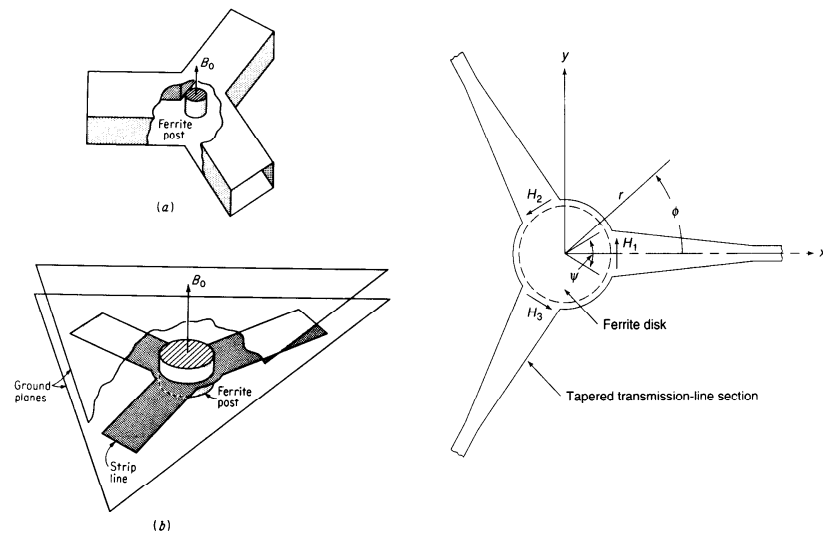
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Four-port circulator



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Three-port circulator

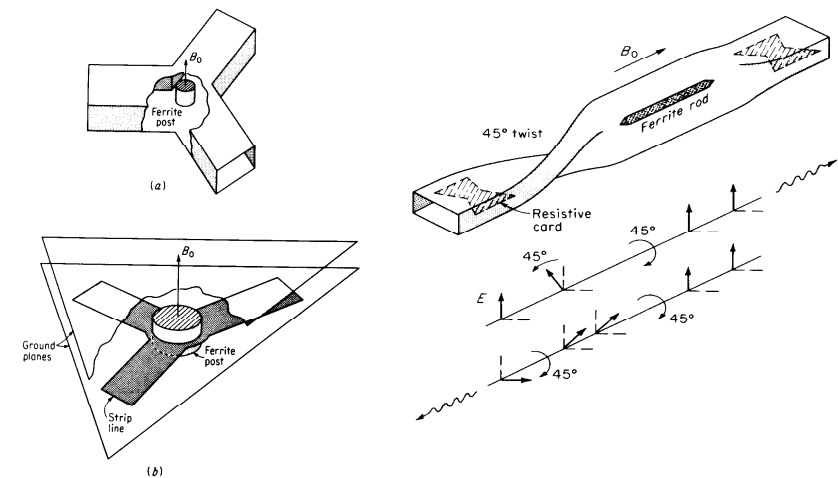


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循環器
circulator

Devices

隔離器
isolator



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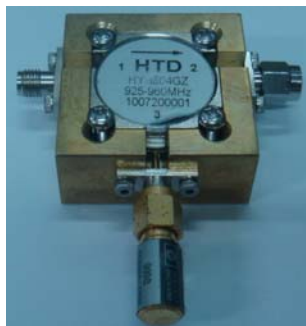
Example : Circulator (I)

Circulator



Fixture

Assembly



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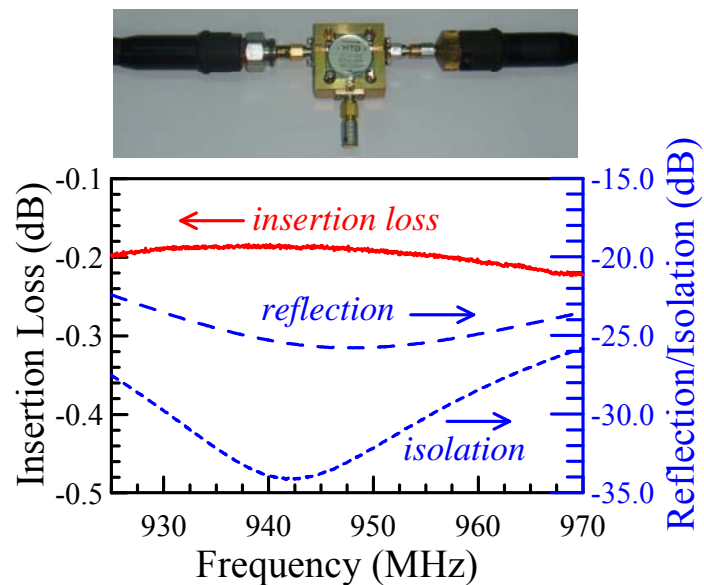
Circulator Measurement (II)

Network Analyzer and Calibration



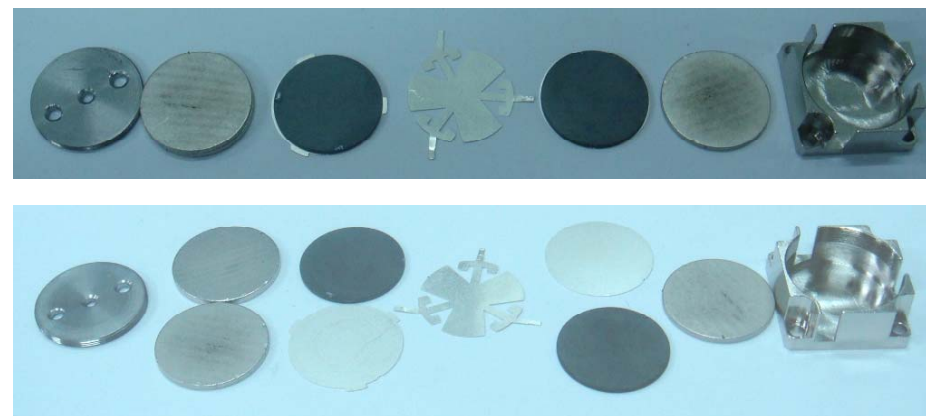
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Circulator Measured Results (III)



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Circulator Construction (IV)

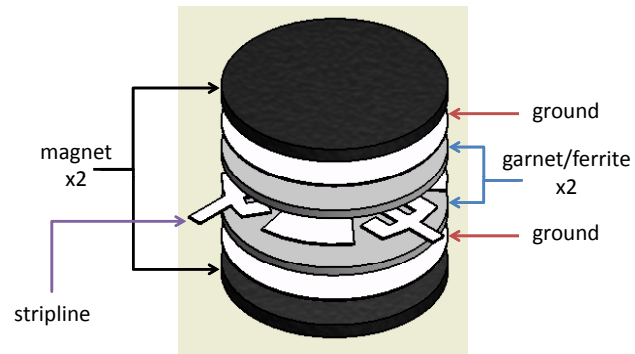


可否再簡化，i.e. cost down?

Ans: 需掌握關鍵技術 → 模擬分析能力。

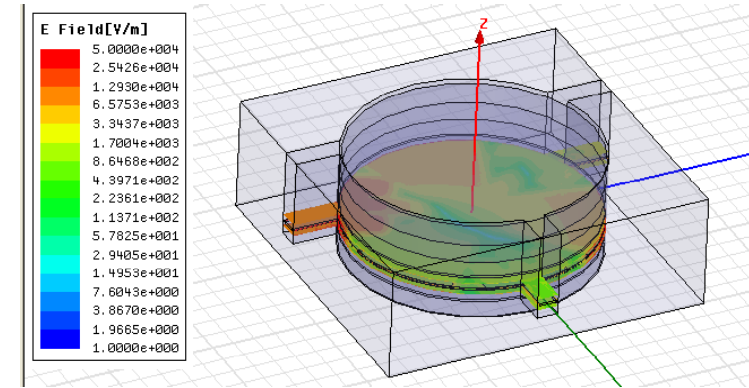
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Circulator Configuration (IV)



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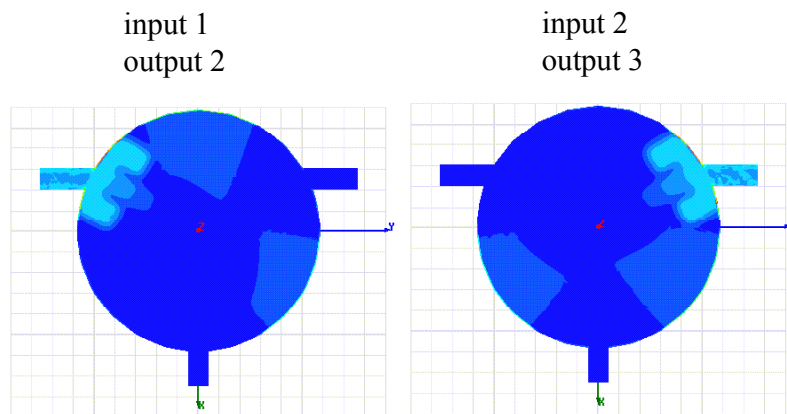
Circulator Simulation (IV)



全尺寸電磁場與熱分析模擬！

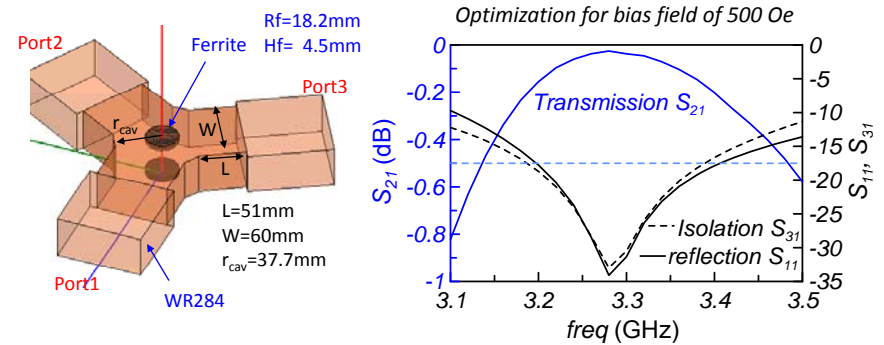
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Circulator Simulated Results (IV)



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Example #2 : S 頻段微波循環器 (波導高功率)




優點: low insertion loss, high isolation, and low reflection

p.s. 本實驗室陳乃慶、蔡育超、蔡鎮鴻設計。

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例二所用Garnet 材料特性比較



Pacific Ceramics, Inc.

Garnets						
Product	Saturation Magnetization 4πMs (gauss)	Dielectric Constant ε'	Loss Tangent tanδ _i (x 10 ⁻⁴)	Resonance Linewidth ΔH @3dB (oe)	Curie Temperature T _c (°C)	
Gadolinium Substituted Yttrium Iron Garnet						
64-1600	1600	15.1	≤ 1	50	280	

Family Type	Saturation Magnetization 4πMs (Gauss)	Line Width ΔH (Gauss @ 3dB)	Dielectric Constant, ε'	Dielectric Loss Tangent Tan δ = ε''/ε' x 10 ⁴	Curie Temperature T _c (°C), Nominal	Spin Wave Line Width ΔH _{sw} (Gauss), Nominal	Remanent Induction B _r (Gauss), Nominal	Coercive Force H _c (Oe), Nominal	α _c ΔM/M _s ΔT x 10 ⁶ , Nominal	Comments (See page 36 for details)
GARNETS – Yttrium Iron Garnet										
YG – 1780 – 45	15.1	≤ 2	280	1.5	—	—	—	2.0	SG	
YG – 1780 – 30	15.1	≤ 2	280	1.5	1275	0.5	2.0	NL		

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Ferrite Circulator for HFSS Simulation

Analytic solution by Eric Chao

優點: low insertion loss, high isolation, and low reflection

$$\begin{aligned}\nabla \cdot \vec{E} &= 0 \\ \nabla \cdot (\hat{u} \cdot \vec{H}) &= 0 \\ \nabla \times \vec{E} &= i\omega \vec{B} = i\omega \hat{u} \cdot \vec{H} \\ \nabla \times \vec{H} &= \frac{\partial \vec{D}}{\partial t} = -i\omega \epsilon \vec{E} \\ \vec{E} &= E_z(\rho, \phi) \hat{e}_z \\ \vec{H} &= H_\rho(\rho, \phi) \hat{e}_\rho + H_\phi(\rho, \phi) \hat{e}_\phi\end{aligned}$$

$$\hat{u} = \begin{pmatrix} \mu & -i\kappa & 0 \\ i\kappa & \mu & 0 \\ 0 & 0 & \mu_0 \end{pmatrix}$$

$$\hat{u} \cdot \vec{H} = \begin{pmatrix} \mu & -i\kappa & 0 \\ i\kappa & \mu & 0 \\ 0 & 0 & \mu_0 \end{pmatrix} \begin{pmatrix} H_\rho \\ H_\phi \\ 0 \end{pmatrix} = \begin{pmatrix} \mu H_\rho - i\kappa H_\phi \\ i\kappa H_\rho + \mu H_\phi \\ 0 \end{pmatrix}$$

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$$\begin{aligned}\nabla \cdot \vec{E} &= \frac{\partial E_z}{\partial z} = 0 \\ \nabla \cdot (\hat{u} \cdot \vec{H}) &= \frac{1}{\rho} \frac{\partial}{\partial \rho} [\rho (\mu H_\rho - i\kappa H_\phi)] + \frac{1}{\rho} \frac{\partial}{\partial \phi} (i\kappa H_\rho + \mu H_\phi) = 0 \\ \nabla \times \vec{E} &= i\omega \hat{u} \cdot \vec{H} \Rightarrow \hat{e}_\rho \frac{1}{\rho} \frac{\partial E_z}{\partial \phi} - \hat{e}_\phi \frac{\partial E_z}{\partial \rho} = i\omega [\hat{e}_\rho (\mu H_\rho - i\kappa H_\phi) + \hat{e}_\phi (i\kappa H_\rho + \mu H_\phi)] \\ \nabla \times \vec{H} &= -i\omega \epsilon \vec{E} \Rightarrow \frac{1}{\rho} \left[\frac{\partial (\rho H_\phi)}{\partial \rho} - \frac{\partial H_\rho}{\partial \phi} \right] = -i\omega \epsilon E_z \\ \begin{cases} H_\rho = -\frac{1}{\omega(\mu^2 - \kappa^2)} \left(\frac{i\mu}{\rho} \frac{\partial E_z}{\partial \phi} + \kappa \frac{\partial E_z}{\partial \rho} \right) \\ H_\phi = \frac{1}{\omega(\mu^2 - \kappa^2)} \left(i\mu \frac{\partial E_z}{\partial \rho} - \frac{\kappa}{\rho} \frac{\partial E_z}{\partial \phi} \right) \end{cases} \Rightarrow \frac{\partial^2 E_z}{\partial \rho^2} + \frac{1}{\rho} \frac{\partial E_z}{\partial \rho} + \frac{1}{\rho^2} \frac{\partial^2 E_z}{\partial \phi^2} + \omega^2 \epsilon \frac{(\mu^2 - \kappa^2)}{\mu} E_z = 0 \\ \frac{1}{\rho} \left[\frac{\partial (\rho H_\phi)}{\partial \rho} - \frac{\partial H_\rho}{\partial \phi} \right] &= -i\omega \epsilon E_z\end{aligned}$$

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$$\begin{aligned}\frac{\partial^2 E_z}{\partial \rho^2} + \frac{1}{\rho} \frac{\partial E_z}{\partial \rho} + \frac{1}{\rho^2} \frac{\partial^2 E_z}{\partial \phi^2} + \omega^2 \epsilon \frac{(\mu^2 - \kappa^2)}{\mu} E_z &= 0 \\ \mu_e = \frac{(\mu^2 - \kappa^2)}{\mu}, \quad \gamma^2 = \omega^2 \epsilon \frac{(\mu^2 - \kappa^2)}{\mu} &= \omega^2 \epsilon \mu_e \\ \frac{\partial^2 E_z}{\partial \rho^2} + \frac{1}{\rho} \frac{\partial E_z}{\partial \rho} + \frac{1}{\rho^2} \frac{\partial^2 E_z}{\partial \phi^2} + \gamma^2 E_z &= 0 \\ E_z &= (A_m e^{im\phi} + B_m e^{-im\phi}) J_m(\gamma_{mn} \rho) \\ H_\rho &= -\frac{1}{\omega(\mu^2 - \kappa^2)} \left(\frac{i\mu}{\rho} \frac{\partial E_z}{\partial \phi} + \kappa \frac{\partial E_z}{\partial \rho} \right) \\ &= -\frac{\kappa \gamma_{mn}}{\omega(\mu^2 - \kappa^2)} \left(\left(J'_m(\gamma_{mn} \rho) - \frac{m\mu}{\kappa \gamma_{mn} \rho} J_m(\gamma_{mn} \rho) \right) A_m e^{im\phi} + \left(J'_m(\gamma_{mn} \rho) + \frac{m\mu}{\kappa \gamma_{mn} \rho} J_m(\gamma_{mn} \rho) \right) B_m e^{-im\phi} \right) \\ H_\phi &= \frac{1}{\omega(\mu^2 - \kappa^2)} \left(i\mu \frac{\partial E_z}{\partial \rho} - \frac{\kappa}{\rho} \frac{\partial E_z}{\partial \phi} \right) \\ &= -\frac{i\mu \gamma_{mn}}{\omega(\mu^2 - \kappa^2)} \left(\left(J'_m(\gamma_{mn} \rho) - \frac{m\kappa}{\mu \gamma_{mn} \rho} J_m(\gamma_{mn} \rho) \right) A_m e^{im\phi} + \left(J'_m(\gamma_{mn} \rho) + \frac{m\kappa}{\mu \gamma_{mn} \rho} J_m(\gamma_{mn} \rho) \right) B_m e^{-im\phi} \right)\end{aligned}$$

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$$E_z = (A_1 e^{i\phi} + B_1 e^{-i\phi}) J_1(\gamma \rho)$$

$$E_z(\rho = R, \phi) = \begin{cases} E_0, & \phi = 0 \\ -E_0, & \phi = \frac{2\pi}{3} \\ 0, & \phi = \frac{4\pi}{3} \end{cases}$$

$$\Rightarrow A_1 = \frac{E_0}{2J_1(\gamma R)} \left(1 + i \frac{1}{\sqrt{3}}\right) \quad \& \quad B_1 = \frac{E_0}{2J_1(\gamma R)} \left(1 - i \frac{1}{\sqrt{3}}\right)$$

$$E_z = (A_1 e^{i\phi} + B_1 e^{-i\phi}) J_1(\gamma \rho) = \left(\frac{E_0}{2J_1(\gamma R)} \left(1 + i \frac{1}{\sqrt{3}}\right) e^{i\phi} + \frac{E_0}{2J_1(\gamma R)} \left(1 - i \frac{1}{\sqrt{3}}\right) e^{-i\phi} \right) J_1(\gamma \rho)$$

$$= \frac{E_0}{J_1(\gamma R)} J_1(\gamma \rho) \left(\cos \phi - \frac{\sin \phi}{\sqrt{3}} \right)$$

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$$B.C. \Rightarrow \hat{n} \times \vec{H} \Big|_{\rho=R} = 0 \Rightarrow -\hat{e}_\rho \times \vec{H} \Big|_{\rho=R} = 0 \Rightarrow H_\phi \Big|_{\rho=R} = 0$$

$$H_\phi \Big|_{\rho=R} = \frac{i\mu\gamma_{mn}}{\omega(\mu^2 - \kappa^2)} \left(\left(J'_m(\gamma_{mn} R) - \frac{m\kappa}{\mu\gamma_{mn} R} J_m(\gamma_{mn} R) \right) A_m e^{im\phi} - \left(J'_m(\gamma_{mn} R) + \frac{m\kappa}{\mu\gamma_{mn} R} J_m(\gamma_{mn} R) \right) B_m e^{-im\phi} \right) = 0$$

$$H_\phi(\rho = R, \phi) = \begin{cases} H_0 & \text{for } -\psi < \phi < \psi \\ H_0 & \text{for } 120^\circ - \psi < \phi < 120^\circ + \psi \\ 0 & \text{elsewhere} \end{cases}$$

$$H_\phi(\rho = R, \phi) = \sum_{m=-\infty}^{\infty} C_m e^{im\phi} = \frac{2H_0\psi}{\pi} + \frac{H_0}{\pi} \sum_{m=1}^{\infty} \frac{\sin(m\psi)}{m} \left(\left(1 + e^{-i\frac{2m\pi}{3}} \right) e^{im\phi} + \left(1 + e^{i\frac{2m\pi}{3}} \right) e^{-im\phi} \right)$$

$$m=1 \Rightarrow H_\phi(\rho = R, \phi) = \frac{H_0 \sin \psi}{2\pi} \left((1 - i\sqrt{3}) e^{i\phi} + (1 + i\sqrt{3}) e^{-i\phi} \right)$$

$$= -\frac{\gamma E_0}{2\sqrt{3}\omega\mu_e J_1(\gamma R)} \left(\left(J'_1(\gamma R) - \frac{\kappa}{\mu\gamma R} J_1(\gamma R) \right) (1 - i\sqrt{3}) e^{i\phi} + \left(J'_1(\gamma R) + \frac{\kappa}{\mu\gamma R} J_1(\gamma R) \right) (1 + i\sqrt{3}) e^{-i\phi} \right)$$

$$\begin{cases} \frac{H_0 \sin \psi}{2\pi} = -\frac{\gamma E_0}{2\sqrt{3}\omega\mu_e J_1(\gamma R)} \left(J'_1(\gamma R) - \frac{\kappa}{\mu\gamma R} J_1(\gamma R) \right) \\ \frac{H_0 \sin \psi}{2\pi} = \frac{\gamma E_0}{2\sqrt{3}\omega\mu_e J_1(\gamma R)} \left(J'_1(\gamma R) + \frac{\kappa}{\mu\gamma R} J_1(\gamma R) \right) \end{cases} \Rightarrow \begin{cases} J'_1(\gamma R) = 0 \\ \frac{H_0 \sin \psi}{\pi} = \frac{E_0}{\sqrt{3}\omega\mu_e \mu R} \frac{\kappa}{H_0} \Rightarrow \frac{E_0}{H_0} = \frac{\sqrt{3}\omega\mu_e \mu R \sin \psi}{\pi\kappa} \end{cases}$$

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$$H_\rho = -\frac{\kappa\gamma_{mn}}{\omega(\mu^2 - \kappa^2)} \left(\left(J'_m(\gamma_{mn} \rho) - \frac{m\mu}{\kappa\gamma_{mn} \rho} J_m(\gamma_{mn} \rho) \right) A_m e^{im\phi} + \left(J'_m(\gamma_{mn} \rho) + \frac{m\mu}{\kappa\gamma_{mn} \rho} J_m(\gamma_{mn} \rho) \right) B_m e^{-im\phi} \right)$$

$$A_1 = \frac{E_0}{2J_1(\gamma R)} \left(1 + i \frac{1}{\sqrt{3}}\right) \quad \& \quad B_1 = \frac{E_0}{2J_1(\gamma R)} \left(1 - i \frac{1}{\sqrt{3}}\right)$$

$$m=1, n=1$$

$$H_\rho = -\frac{\kappa\gamma_{11} E_0}{\omega(\mu^2 - \kappa^2)} \left(\left(J'_1(\gamma_{11} \rho) - \frac{\mu}{\kappa\gamma_{11} \rho} J_1(\gamma_{11} \rho) \right) A_1 e^{i\phi} + \left(J'_1(\gamma_{11} \rho) + \frac{\mu}{\kappa\gamma_{11} \rho} J_1(\gamma_{11} \rho) \right) B_1 e^{-i\phi} \right)$$

$$= -\frac{\kappa\gamma_{11} E_0}{2\omega(\mu^2 - \kappa^2) J_1(\gamma_{11} R)} \left(\left(J'_1(\gamma_{11} \rho) - \frac{\mu}{\kappa\gamma_{11} \rho} J_1(\gamma_{11} \rho) \right) \left(1 + i \frac{1}{\sqrt{3}} \right) e^{i\phi} + \left(J'_1(\gamma_{11} \rho) + \frac{\mu}{\kappa\gamma_{11} \rho} J_1(\gamma_{11} \rho) \right) \left(1 - i \frac{1}{\sqrt{3}} \right) e^{-i\phi} \right)$$

$$= -\frac{\kappa\gamma_{11} E_0}{\omega(\mu^2 - \kappa^2) J_1(\gamma_{11} R)} \left(\left(J'_1(\gamma_{11} \rho) - i \frac{\mu}{\kappa\gamma_{11} \rho} \frac{J_1(\gamma_{11} \rho)}{\sqrt{3}} \right) \cos \phi - \left(\frac{J'_1(\gamma_{11} \rho)}{\sqrt{3}} + \frac{i\mu}{\kappa\gamma_{11} \rho} J_1(\gamma_{11} \rho) \right) \sin \phi \right)$$

$$H_\rho(\rho = R, \phi) = -\frac{\kappa\gamma_{mn}}{\omega(\mu^2 - \kappa^2)} \left((-J_m(\gamma_{mn} R)) A_m e^{im\phi} + \left(\frac{m\mu}{\kappa\gamma_{mn} R} J_m(\gamma_{mn} R) \right) B_m e^{-im\phi} \right)$$

$$= \frac{m\mu}{\omega R(\mu^2 - \kappa^2)} (A_m e^{im\phi} - B_m e^{-im\phi}) J_m(\gamma_{mn} R) = \frac{\mu}{\omega R(\mu^2 - \kappa^2)} (A_1 e^{i\phi} - B_1 e^{-i\phi}) J_1(\gamma'_{11})$$

$$= \frac{\mu E_0}{2\omega R(\mu^2 - \kappa^2)} \left(\left(1 + i \frac{1}{\sqrt{3}} \right) e^{i\phi} - \left(1 - i \frac{1}{\sqrt{3}} \right) e^{-i\phi} \right) = \frac{iH_0}{x'_{11}} \left(\sin \phi + \frac{1}{\sqrt{3}} \cos \phi \right)$$

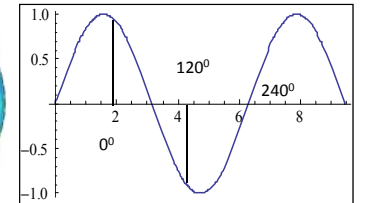
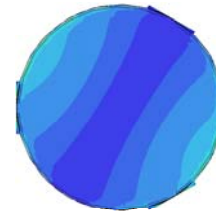
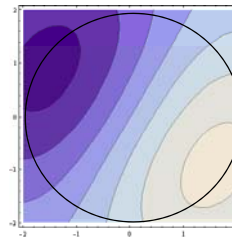
39

$$E_z = \frac{E_0}{J_1(\gamma R)} J_1(\gamma \rho) \left(\cos \phi - \frac{\sin \phi}{\sqrt{3}} \right) = \frac{E_0}{J_1(x'_{11})} J_1\left(\frac{x'_{11} \rho}{R}\right) \left(\cos \phi - \frac{\sin \phi}{\sqrt{3}} \right)$$

$$\begin{cases} J'_1(\gamma R) = 0 \Rightarrow \boxed{\gamma_{11} R = x'_{11} = 1.841 = \omega_{110} \sqrt{\epsilon \mu_e} R} \\ \frac{H_0 \sin \psi}{\pi} = \frac{E_0}{\sqrt{3}\omega\mu_e \mu R} \frac{\kappa}{H_0} \Rightarrow \frac{E_0}{H_0} = \frac{\sqrt{3}\omega_{110} \mu_e \mu R \sin \psi}{\pi\kappa} = \frac{\sqrt{3}\mu_e \mu \sin \psi}{\pi\kappa} \frac{1.841}{\sqrt{\epsilon \mu_e}} \end{cases}$$

$$\frac{E_0}{H_0} = \frac{\sqrt{3}\omega_{110} \mu_e \mu R \sin \psi}{\pi\kappa} = \sqrt{\frac{3\mu_e}{\epsilon}} \frac{1.841 \mu \sin \psi}{\pi\kappa} = \sqrt{\frac{\mu_e}{\epsilon}} \Rightarrow \boxed{\sin \psi = \frac{\pi}{1.841\sqrt{3}} \left(\frac{\kappa}{\mu} \right)}$$

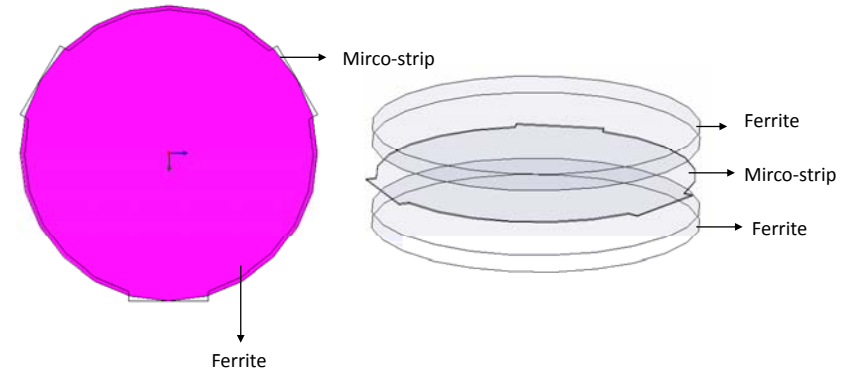
$$\sin \psi = \frac{\pi}{1.841\sqrt{3}} \left(\frac{\kappa}{\mu} \right) = \frac{\pi}{1.841\sqrt{3}} \left(\frac{\omega\omega_m}{\omega_0^2 - \omega^2 + \omega_0\omega_m} \right)$$



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HSS Simulation

Simulation Model



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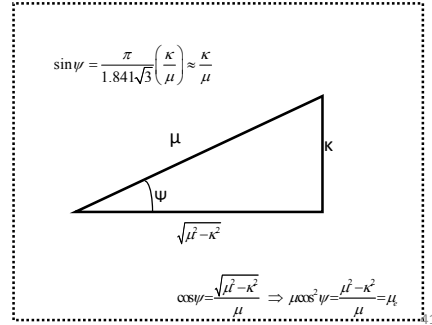
$$H_\phi(\rho, \phi) = -\frac{\gamma E_0}{2\sqrt{3}\omega\mu_e J_1(x'_{11})} \left(\left(J'_1(\gamma\rho) - \frac{\kappa}{\mu\gamma\rho} J_1(\gamma\rho) \right) (1-i\sqrt{3})e^{i\phi} + \left(J'_1(\gamma\rho) + \frac{\kappa}{\mu\gamma\rho} J_1(\gamma\rho) \right) (1+i\sqrt{3})e^{-i\phi} \right)$$

$$x = \gamma\rho \Rightarrow \begin{cases} J'_1(x) - \frac{\kappa}{\mu x} J_1(x) = 0 \\ J'_1(x) + \frac{\kappa}{\mu x} J_1(x) = 0 \end{cases} \Rightarrow \begin{cases} J'_1(x^-) - \frac{\kappa}{\mu x^-} J_1(x^-) = 0 \\ J'_1(x^+) + \frac{\kappa}{\mu x^+} J_1(x^+) = 0 \end{cases}$$

$$\Rightarrow x'_{11} = \gamma_{11} R = \sqrt{\omega_{110}^2 \epsilon \frac{(\mu^2 - \kappa^2)}{\mu}} R = \omega_{110} \sqrt{\epsilon \mu_e} R$$

$$R = \frac{x'_{11}}{\omega_{110} \sqrt{\epsilon \mu_e}}$$

$$\omega_{110}^+ = \frac{x'_{11}^+}{R \sqrt{\epsilon \mu_e}} \quad \& \quad \omega_{110}^- = \frac{x'_{11}^-}{R \sqrt{\epsilon \mu_e}}$$



Example 1 :

$$R = \frac{x'_{11}}{\omega_{110} \sqrt{\epsilon \mu_e}} = 19.843 \text{ mm}$$

$$f_0 = \omega_0 / 2\pi = \mu_0 \gamma H_0 / 2\pi = (2.8 \text{ MHz} / \text{Oe}) \times (H_0 \text{ Oe}) = 10.82 \text{ GHz}$$

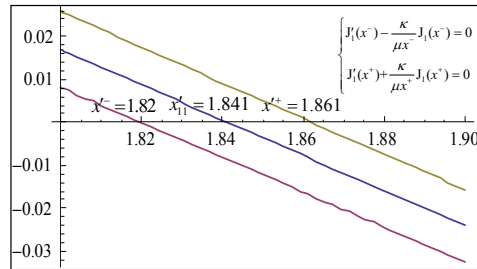
$$f_m = \omega_m / 2\pi = \mu_0 \gamma M_s / 2\pi = (2.8 \text{ MHz} / \text{Oe}) \times (4\pi M_s \text{ Gauss}) = 4.9 \text{ GHz}$$

$$f = 0.945 \text{ GHz}$$

$$\mu = \mu_0 \left(1 + \frac{\omega_0 \omega_m}{\omega_0^2 - \omega^2} \right) = 1.4563 \mu_0$$

$$\kappa = \mu_0 \frac{\omega \omega_m}{\omega_0^2 - \omega^2} = 0.03985 \mu_0$$

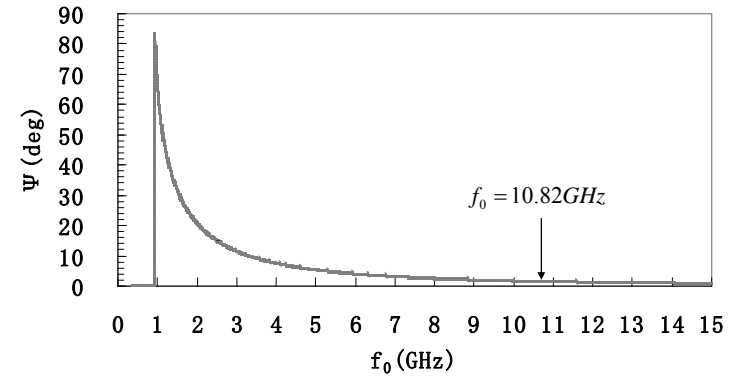
$$\mu_e = \frac{(\mu^2 - \kappa^2)}{\mu} = 1.4552 \mu_0$$



$$\omega_{110}^+ = \frac{x'_{11}^+}{R \sqrt{\epsilon \mu_e}} = \frac{x'_{11}^+}{x'_{11}} \omega_{110} \Rightarrow f_{110}^+ = \frac{x'_{11}^+}{x'_{11}} f_{110} = \frac{1.863}{1.841} \times 0.945 \text{ GHz} = 0.956 \text{ GHz}$$

$$\omega_{110}^- = \frac{x'_{11}^-}{R \sqrt{\epsilon \mu_e}} = \frac{x'_{11}^-}{x'_{11}} \omega_{110} \Rightarrow f_{110}^- = \frac{x'_{11}^-}{x'_{11}} f_{110} = \frac{1.820}{1.841} \times 0.945 \text{ GHz} = 0.934 \text{ GHz}$$

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$$\sin \psi = \frac{\pi}{1.841\sqrt{3}} \left(\frac{f \times f_m}{f_0^2 - f^2 + f_0 f_m} \right) \Rightarrow f = 0.945 \text{ GHz} \quad \& \quad f_m = 4.9 \text{ GHz}$$

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Material Name
YG-1780-45

Properties of the Material

Name	Type	Value	Units
Relative Permittivity	Simple	15.1	
Relative Permeability	Simple	1.4563	
Bulk Conductivity	Simple	0	siemens/m
Dielectric Loss Tangent	Simple	0	
Magnetic Loss Tangent	Simple	0	
Magnetic Saturation	Simple	1780	gauss
Landé G Factor	Simple	2	
Delta H	Simple	45	De
Measured Frequency	Simple	9.4e+009	Hz

View/Edit Material for
☒ Active Design
☐ This Product
☐ All Products

View/Edit Modifier for
☐ Thermal Modifier

Validate Material

Set Frequency Dependency ... Calculate Properties for ...

Reset OK Cancel

Project Manager

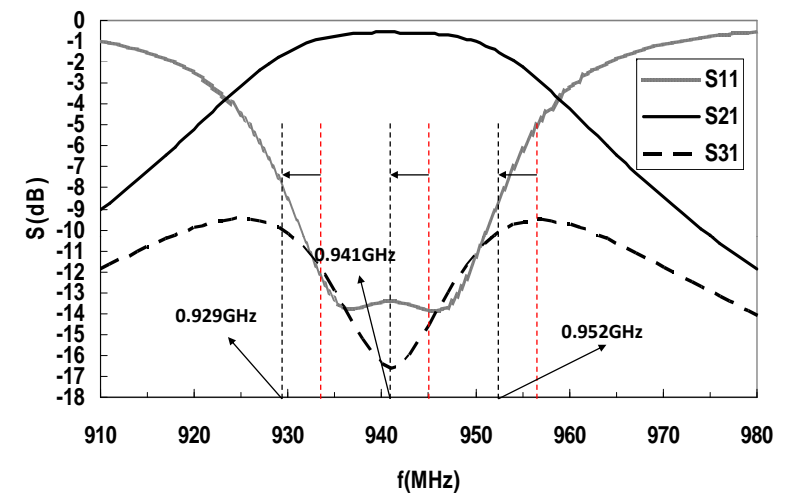
MAXELL-1780
 HFSS Design1
 Definitions

Project

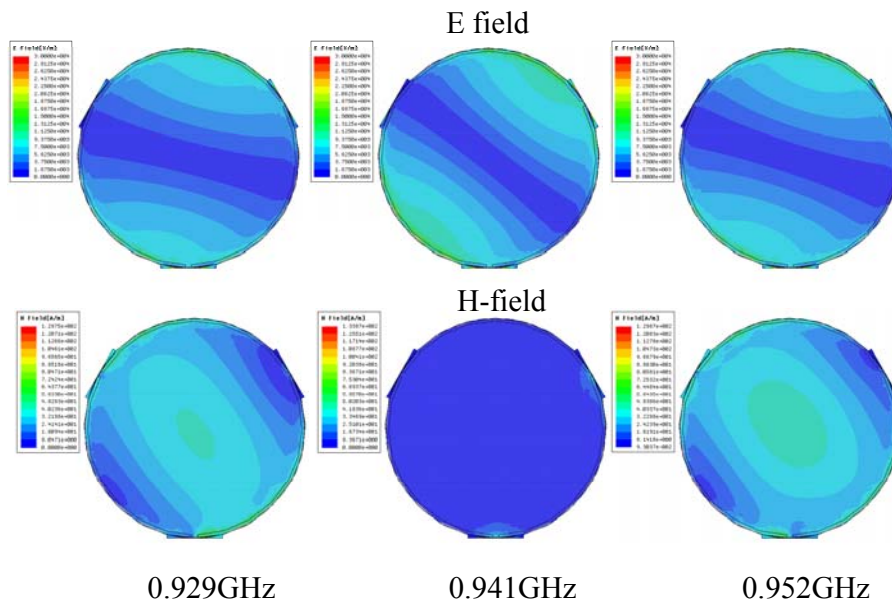
Properties

Name	Value	Unit	Evaluate
H0	307500		307500
RC	19.3	mm	19.3mm
w/1	1.051	mm	1.051mm
D	0.1	mm	0.1mm
H	1.5	mm	1.5mm
w/	10	mm	10mm
T	0.001	mm	0.001mm
RR	19.843	mm	19.843mm
Angle	1.55	deg	1.55deg

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Example 2 :

$$R = \frac{x'_{11}}{\omega_{110} \sqrt{\epsilon \mu_e}} = 22.024 \text{ mm}$$

$$f_0 = \omega_0 / 2\pi = \mu_0 \gamma H_0 / 2\pi = (2.8 \text{ MHz} / \text{Oe}) \times (H_0 \text{ Oe}) = 5.65 \text{ GHz}$$

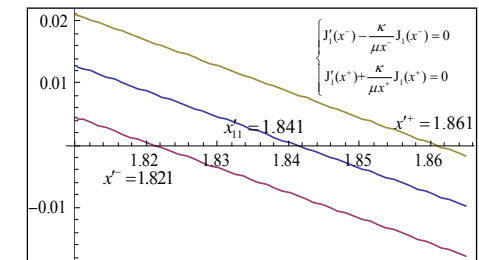
$$f_m = \omega_m / 2\pi = \mu_0 \gamma M_s / 2\pi = (2.8 \text{ MHz} / \text{Oe}) \times (4\pi M_s \text{ Gauss}) = 1 \text{ GHz}$$

$$f = 0.945 \text{ GHz}$$

$$\mu = \mu_0 \left(1 + \frac{\omega_0 \omega_m}{\omega_0^2 - \omega^2} \right) = 1.1821 \mu_0$$

$$\kappa = \mu_0 \frac{\omega \omega_m}{\omega_0^2 - \omega^2} = 0.030455 \mu_0$$

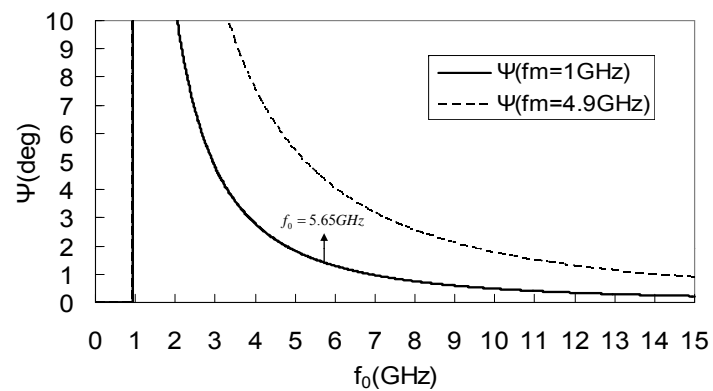
$$\mu_e = \frac{(\mu^2 - \kappa^2)}{\mu} = 1.1813 \mu_0$$



$$\omega_{110}^+ = \frac{x'_{11}^+}{R \sqrt{\epsilon \mu_e}} = \frac{x'_{11}^+}{x'_{11}} \omega_{110} \Rightarrow f_{110}^+ = \frac{x'_{11}^+}{x'_{11}} f_{110} = \frac{1.861}{1.841} \times 0.945 \text{ GHz} = 0.955 \text{ GHz}$$

$$\omega_{110}^- = \frac{x'_{11}^-}{R \sqrt{\epsilon \mu_e}} = \frac{x'_{11}^-}{x'_{11}} \omega_{110} \Rightarrow f_{110}^- = \frac{x'_{11}^-}{x'_{11}} f_{110} = \frac{1.821}{1.841} \times 0.945 \text{ GHz} = 0.935 \text{ GHz}$$

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$$\sin \psi = \frac{\pi}{1.841\sqrt{3}} \left(\frac{f \times f_m}{f_0^2 - f^2 + f_0 f_m} \right) \Rightarrow f = 0.945\text{GHz} \quad \& \quad f_m = 1\text{GHz}$$

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Material Name: YG-357

Properties of the Material:

Name	Type	Value	Units
Relative Permittivity	Simple	15.1	
Relative Permeability	Simple	1.181	
Bulk Conductivity	Simple	0	siemens/m
Dielectric Loss Tangent	Simple	0	
Magnetic Loss Tangent	Simple	0	
Magnetic Saturation	Simple	357	gauss
Lande G Factor	Simple	2	
Delta H	Simple	45	De
Measured Frequency	Simple	9.4e+009	Hz

View/Edit Material for:

- ☒ Active Design
- ☐ This Product
- ☐ All Products

View/Edit Modifier for:

- ☐ Thermal Modifier

Validate Material

Set Frequency Dependency... Calculate Properties for:

Reset OK Cancel

MAXWELL-1

HFSSTDesign1 (DrivenModel)

Definitions

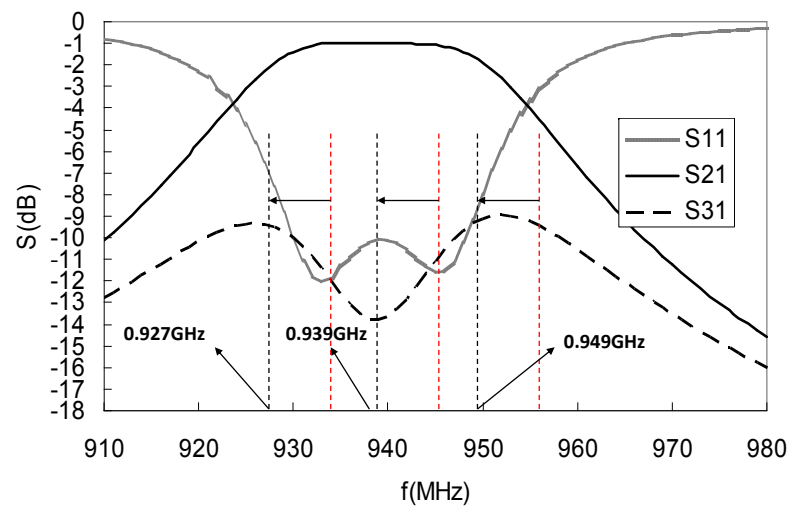
Project

Properties

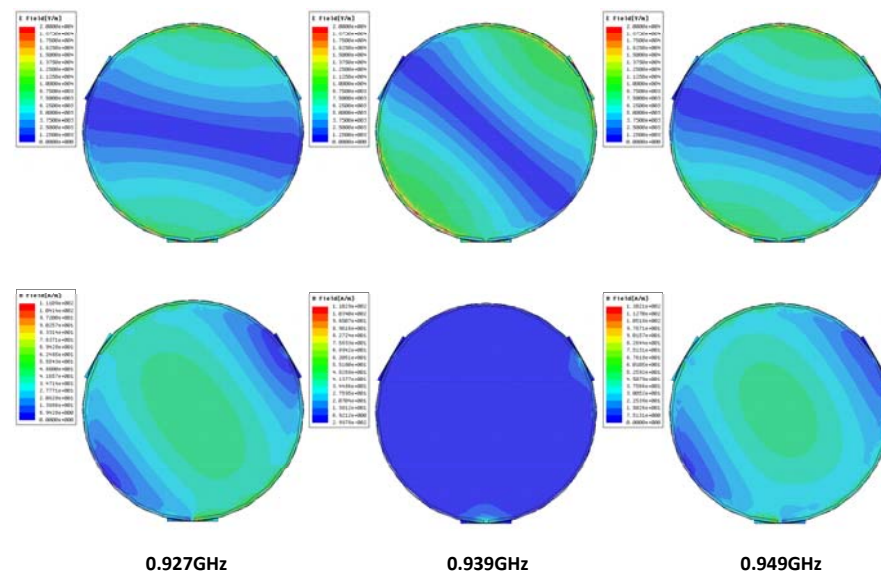
Name	Value	Unit	Evaluate
H0	160576		160576
RC	21.475	mm	21.475mm
W1	1	mm	1mm
D	0.1	mm	0.1mm
H	1.5	mm	1.5mm
W	10	mm	10mm
T	0.001	mm	0.001mm
RR	22.023	mm	22.023mm
Angle	1.55	deg	1.55deg

Variables

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